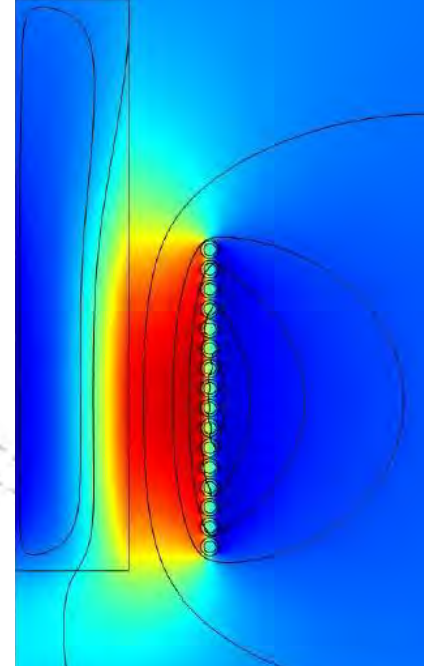




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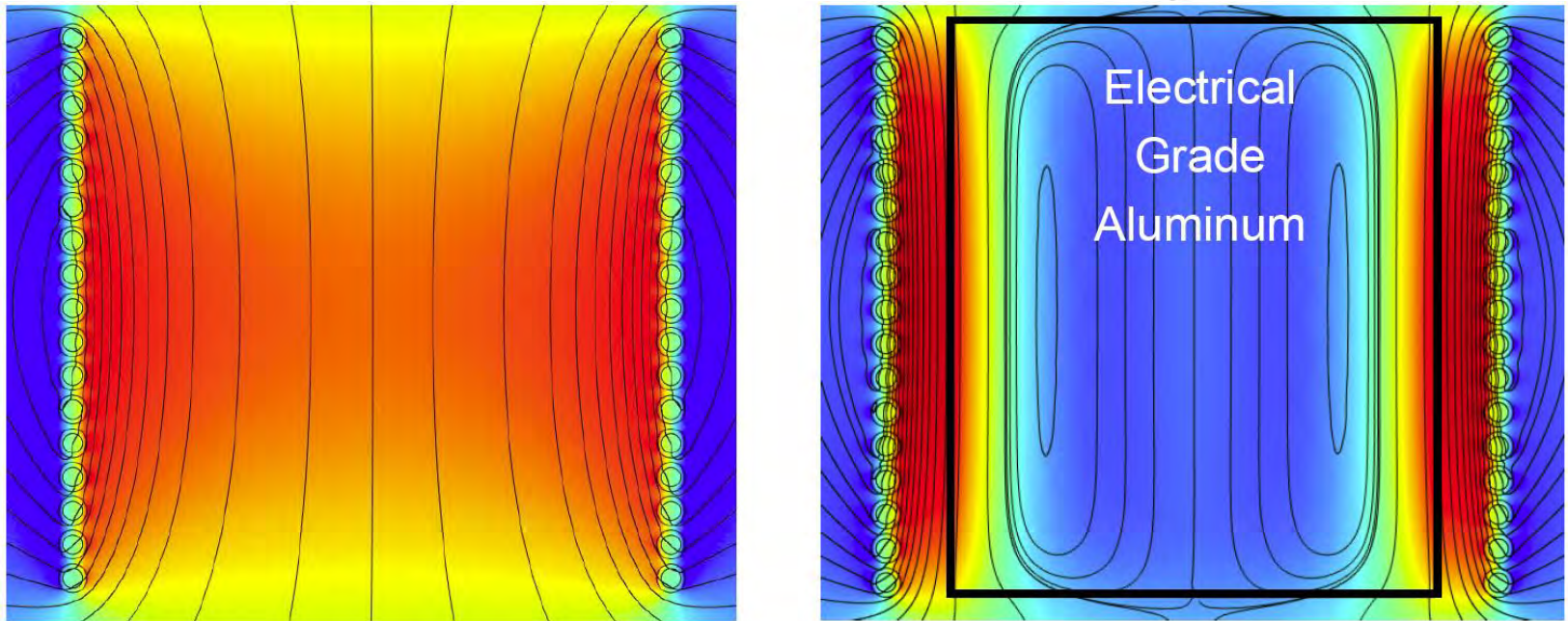
Analytical and Experimental Validation of Electromagnetic Simulations Using COMSOL[®], re Inductance, Induction Heating and Magnetic Fields

Presented by M.W. Kennedy

Co-Authors: S. Akhtar, J.A. Bakken, R.E. Aune



Impact of work piece on magnetic field distribution and relative strength, 50 Hz



Infinite coil:

$$|\vec{B}_\infty| = \frac{\mu_o \mu_r N_c I_c}{l_c}$$

Short coil:

$$|\vec{B}_o| = k_N |\vec{B}_\infty|$$

Coil with a work piece:

$$k_N^* = k_N \left(1 - \left(\frac{D_w - \delta_w}{D_c + \delta_c} \right)^2 \right) + \left(\frac{D_w - \delta_w}{D_c + \delta_c} \right)^2$$

Short coils have non-uniform fields both axially and radially. The field is strongly influenced by the work piece/coil geometry and the electromagnetic penetration depth in the work piece.

Modelling of magnetic fields and induction heating with COMSOL[®]

- Should the coil be voltage or current driven?
- How big does the magnetic domain need to be to simulate an infinite external volume?
i.e. when is the coil flux density estimated with 100% accuracy for a given applied magneto-motive force (NI)?

Modelling of magnetic fields and induction heating with COMSOL[®]

- Which domain, “single-turn” or “multi-turn” can be used and under what circumstances?
- What mesh is required to obtain accurate results at different frequencies? How do we relate this to the physics?
- How accurately can a 2D axial symmetric model estimate magnetic fields and heating rates for cylindrical work pieces in experimental helical coils?

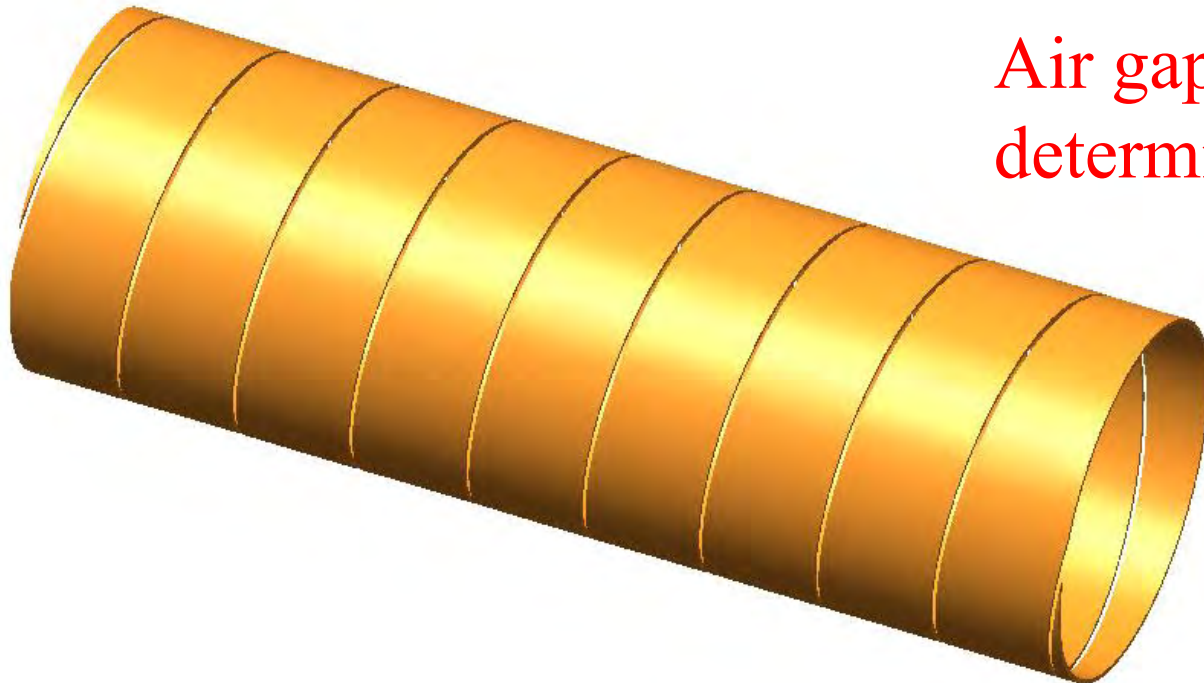
How to find the required magnetic domain size? 5

$$L_o = \frac{8\mu_0 N^2 r^3}{3\ell^2} \left[\frac{2k^2 - 1}{k^3} E(k) + \frac{1 - k^2}{k^3} K(k) - 1 \right] \quad k = \sqrt{\frac{4r^2}{4r^2 + \ell^2}}$$

$$L_o = \frac{k_N A_c N_c |\vec{B}_\infty|}{I_c}$$

k_N = Nagaoka short coil correction factor. Can be solved to double precision accuracy.

**Air gap flux density²
determines the heating rate!**



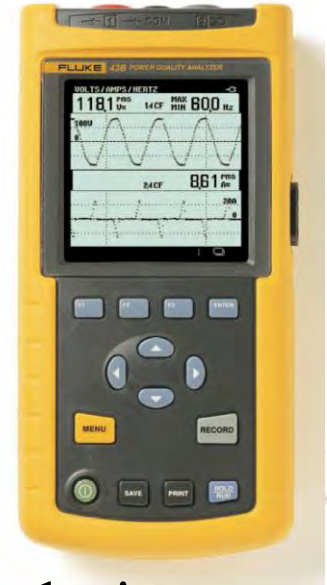
Comparison of COMSOL[®] and analytical inductance of a current sheet

Ratio of Magnetic Domain Dimensions to Coil Dimensions	COMSOL Calculated Inductance (μH)	COMSOL - Analytical Solution Difference (%)
2.00	22.7563	-13.82
4.00	25.9502	-1.72
6.00	26.2783	-0.48
10.00	26.3870	-0.07
14.00	26.4057	0.00
20.00	26.4129	0.03

Error in inductance is the same as for the flux density and is then squared when calculating heating rate!

Theoretical answer = 26.4051 μH .
Ratio of 14 gives ideal results.

Induction heating instrumentation



Magnetic field measurements
Axial/Transverse
From 0.1 μ T-30T
+/- 1.0% AC
Standards from 500-2000 Gauss

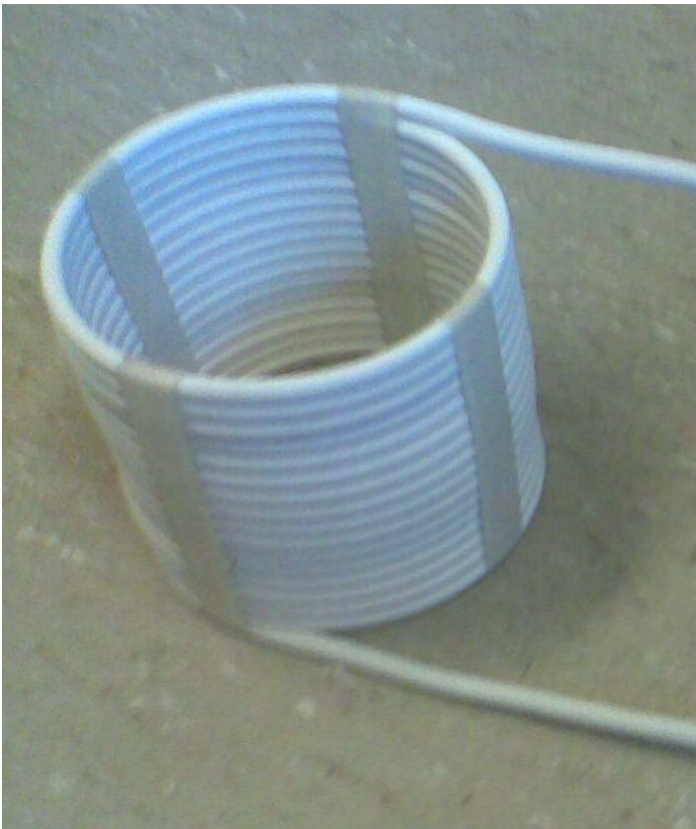
Electrical analysis:

1. V, I, P (+/-100 W), p.f.
2. Inductance
3. Harmonics
4. Current +/- 1% (usable up to 100 kHz)



Electrical conductivity accuracy of +/- 0.5%
Standards +/- 0.01% IACS

Coil and work piece



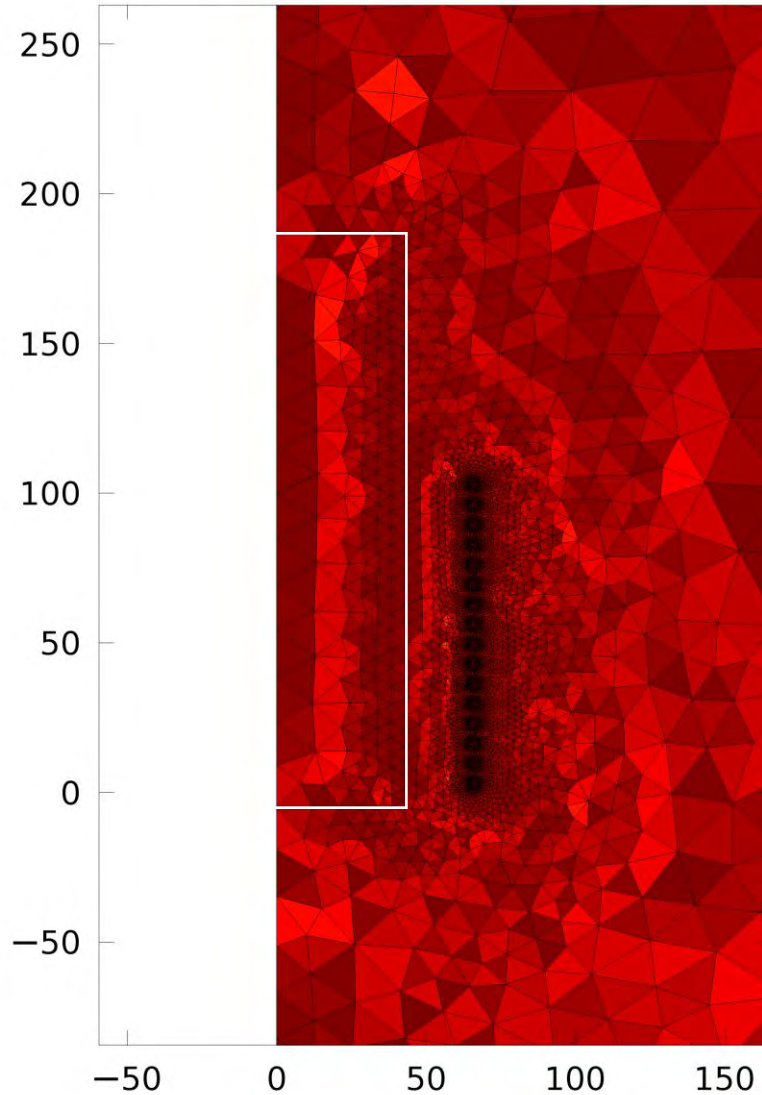
6 mm by 1 mm wall, DHP copper
80% IACS electrical conductivity
Insulated with glass fibre sleeves



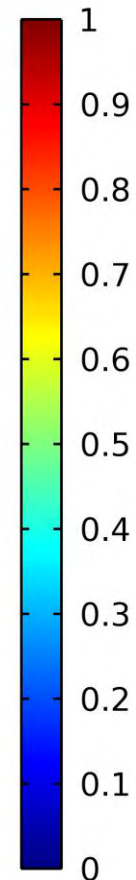
A356 aluminium alloy
48% IACS conductivity

Mesh 1

freq(1)=50 Mesh: Quality



▲ 1



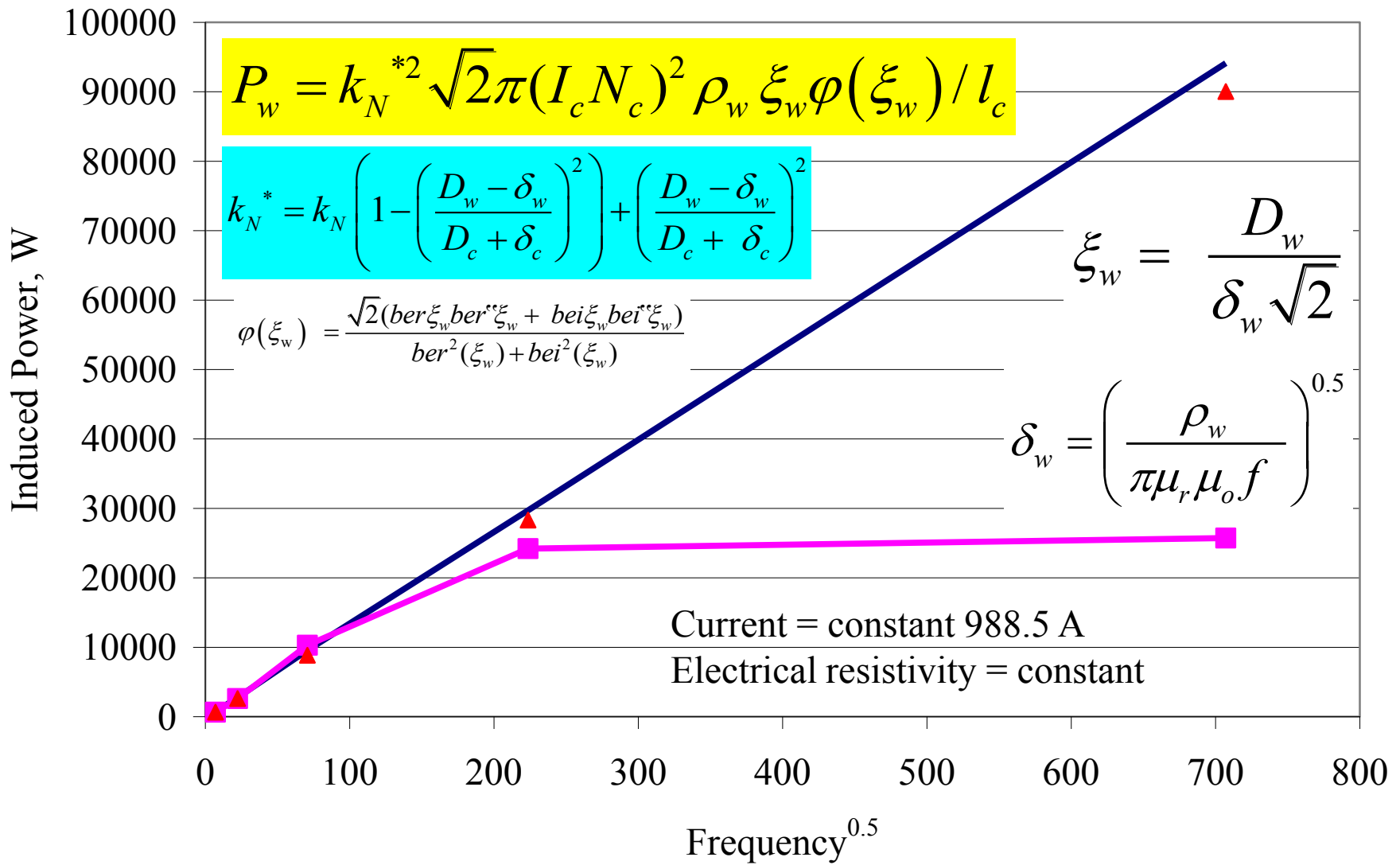
▼ 0.8184

Induction heating using mesh 1

Frequency (Hz)	Experimental Power (W)	Analytical Power (W)	Mesh 1 Power (W)	Mesh 1-Analytical Difference (%)	δ (mm)
50	696	691	650	-6.0	14.50
500	N/A	2768	2604	-5.9	4.59
5000	N/A	9549	10280	7.7	1.45
50000	N/A	29697	24211	-18.5	0.46
500000	N/A	94123	25728	-72.7	0.14
Mesh 1 spacing at work piece interface =					5.10

At „High Frequency“ the power induced should change by \sqrt{f} . Also the first electromagnetic penetration depth will contain 63% of the total current and 86% of the power, with an **exponential gradient squared**.

Variation of heating rate with frequency^{0.5} 11



— Analytical solution —■ COMSOL mesh 1 —▲ COMSOL mesh 2

Induction heating using boundary meshes

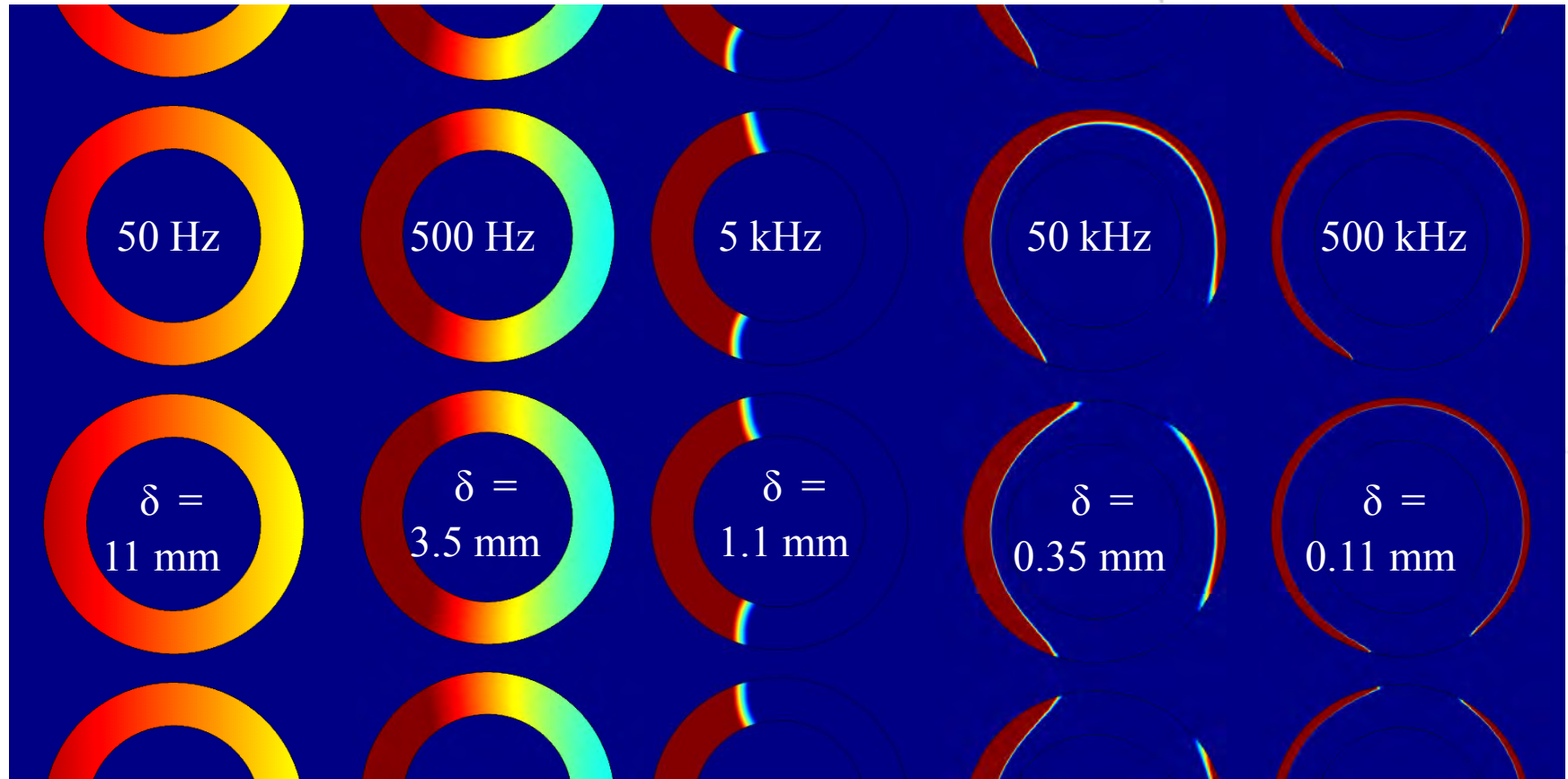
Frequency (Hz)	Experimental Power (W)	Analytical Power (W)	Mesh 2 Power (W)	Mesh 2- Analytical Difference (%)	δ (mm)
50	696	691	650	-6.0	14.5
500	N/A	2768	2597	-6.2	4.59
5000	N/A	9549	8834	-7.5	1.45
50000	N/A	29697	28305	-4.7	0.46
500000	N/A	94123	90029	-4.3	0.14
Mesh 2 spacing at work piece interface =					0.02

Boundary meshes allow accurate calculation to extremely high frequency. Mesh spacing should be $< \delta$.

„Single-turn“ vs. „Multi-turn“ domain

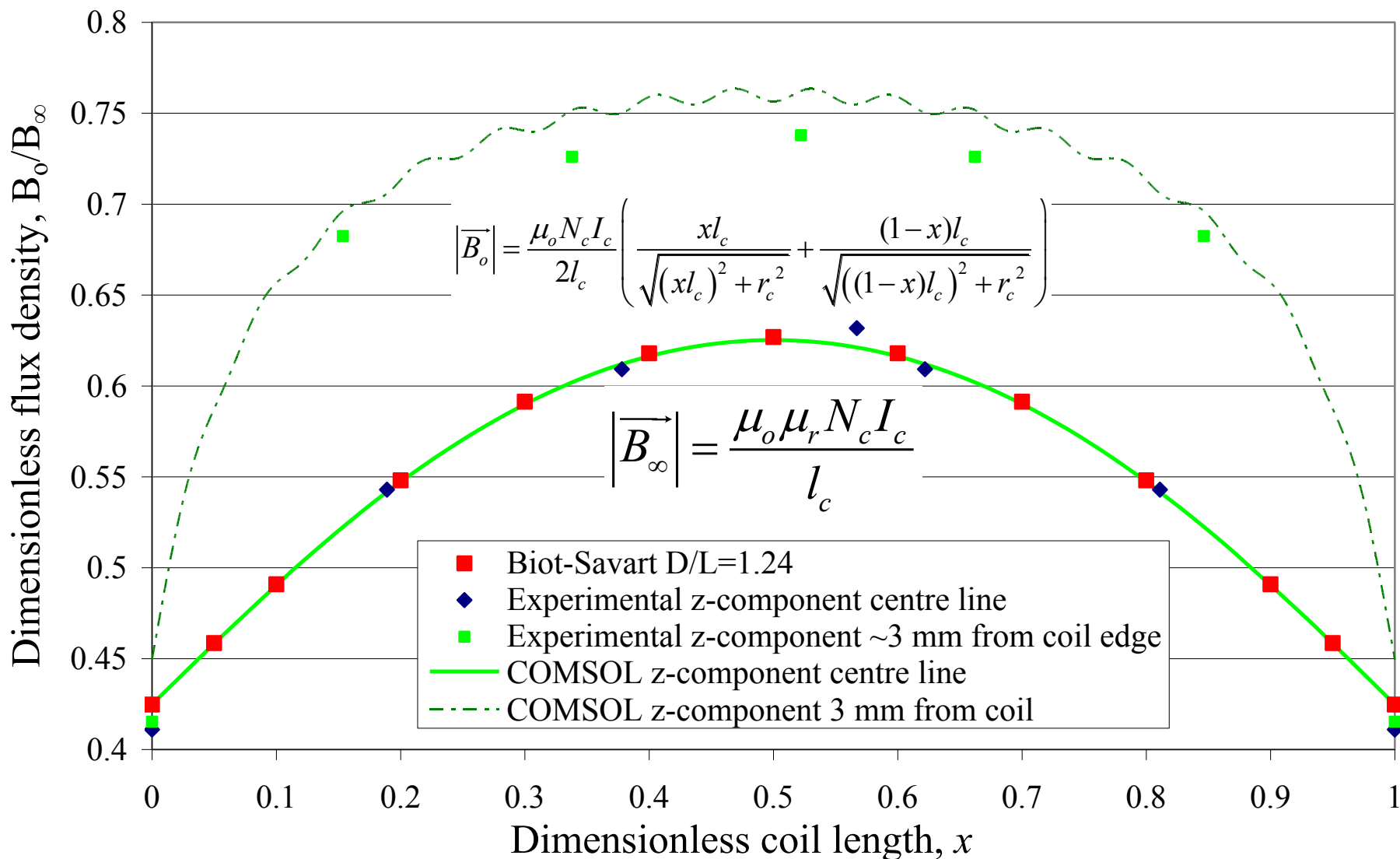
13

← To centreline of coil

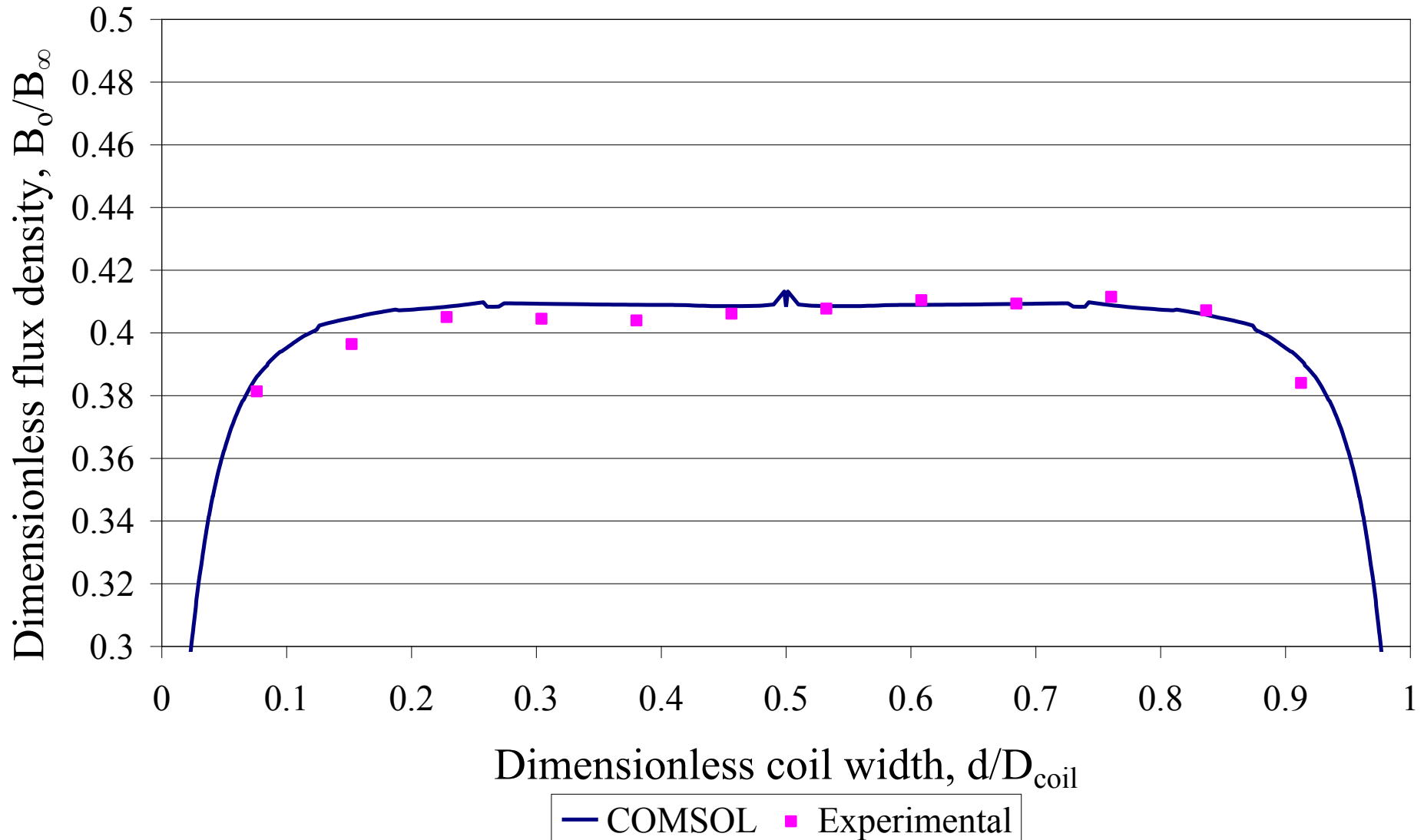


If $\delta <$ tubing diameter, must use “Single-turn” domain and ideally a current driven coil, voltage driven „Multi-turn“ results will be wrong.

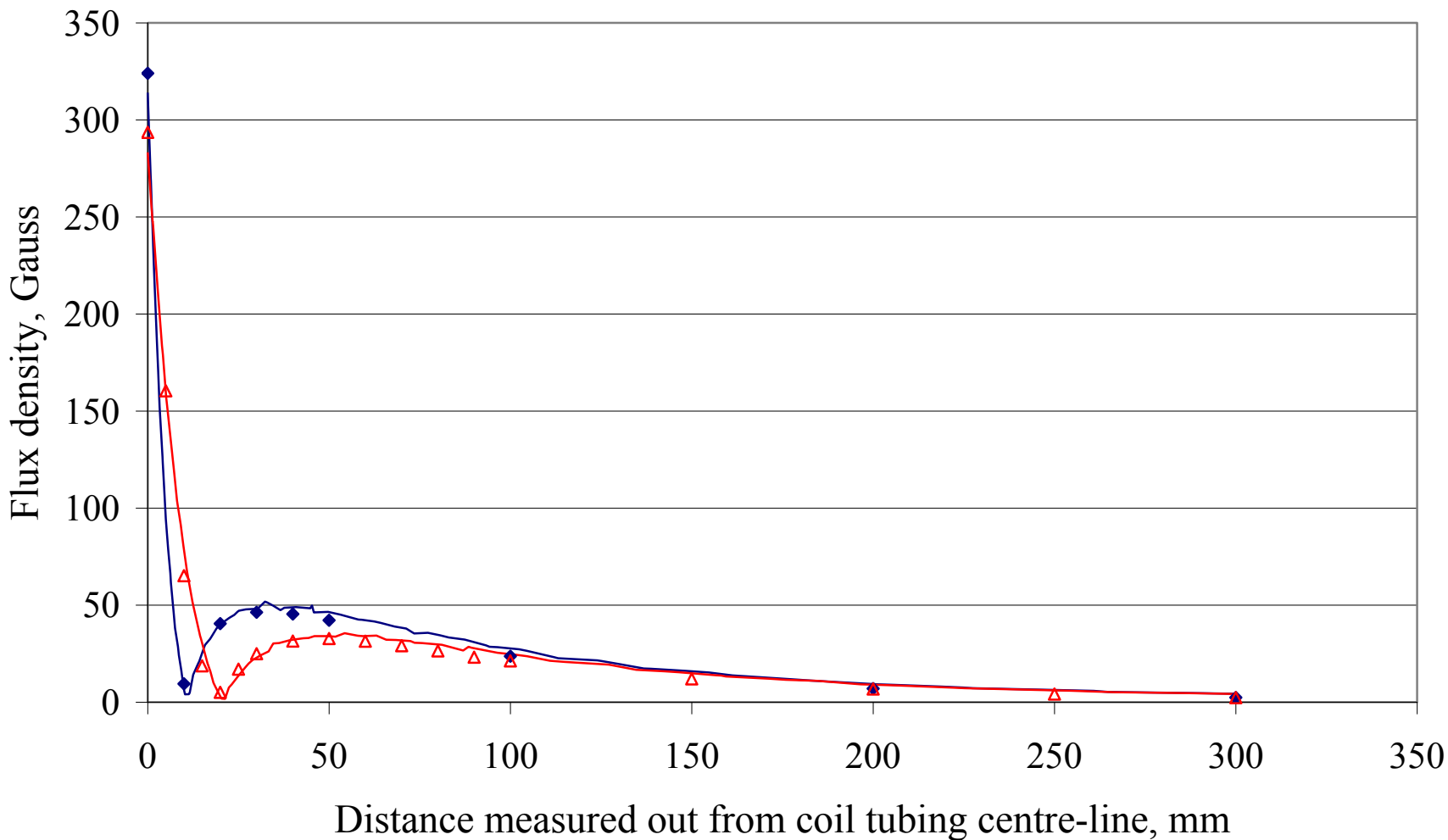
Dimensionless flux density vs. position



Transverse probe magnetic field measurements 2 mm over top of short coil



Transverse probe magnetic field measurements from side of short coil



◆ Experimental -5mm — COMSOL-5mm ▲ Experimental -10mm — COMSOL-10mm

Conclusions

- Current driven coils are recommended, they give the correct magneto-motive-force (NI).
- 14 times the coil size is sufficient to simulate an infinite external volume, for 2D axial symmetric models.
- „Single-turn“ domain is recommended, it gives correct results at all frequencies.

Conclusions

- „Multi-turn“ domain can be used if the electromagnetic penetration depth is greater than the coil tubing diameter.
- Due to the extremely steep current gradients at the surface of the work piece at high frequencies (small δ), boundary meshes should be used to give a mesh spacing $< \delta$.
- Magnetic field estimates with error $< 1-2\%$ and heating estimates with errors $< 6\%$ can be obtained. (**Note:** New calorific measurements have verified errors to be $< 2\%$!)

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Thank you for your attention !

Mark William Kennedy

Department of Materials Science and Engineering

Phone: +47 73595164

Mobile: +47 92219891

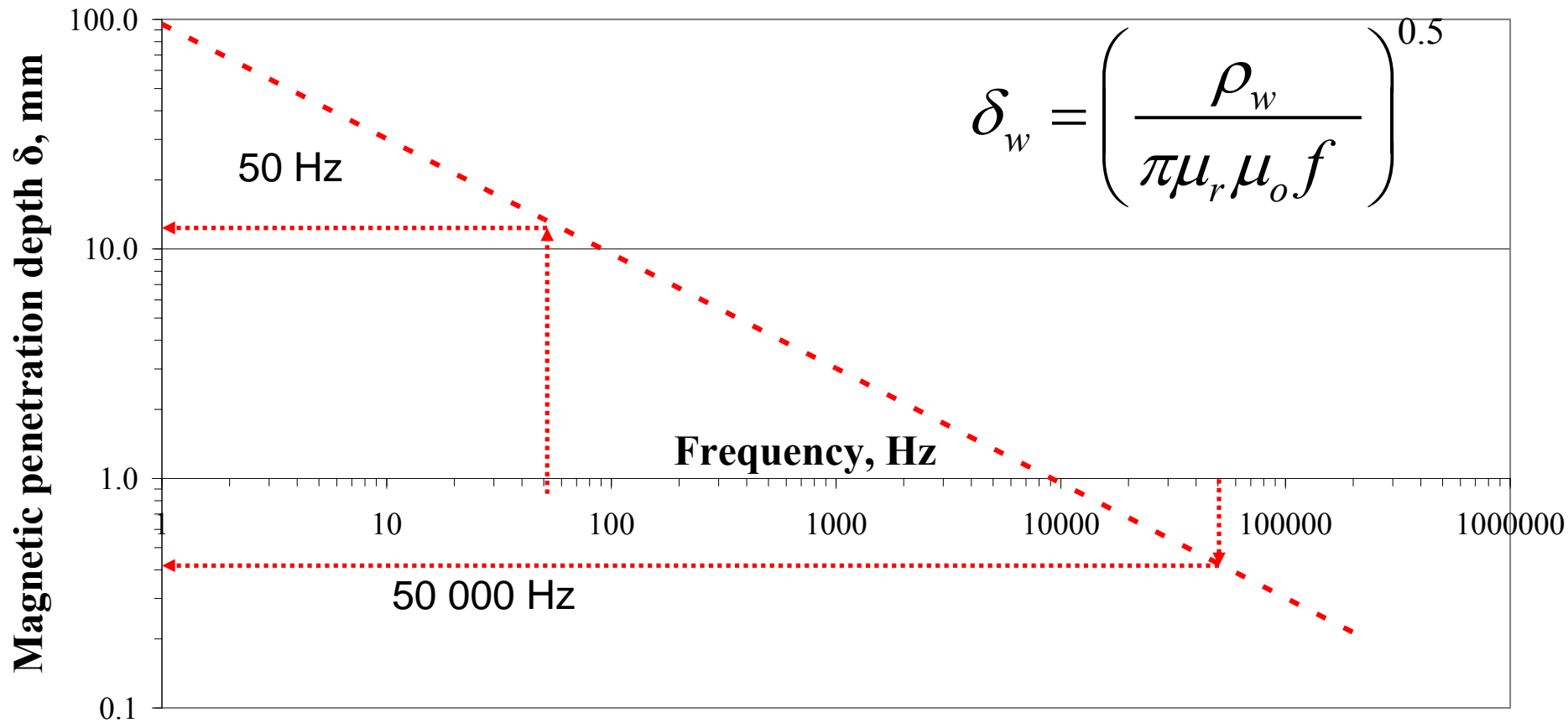
E-mail: mark.kennedy@material.ntnu.no



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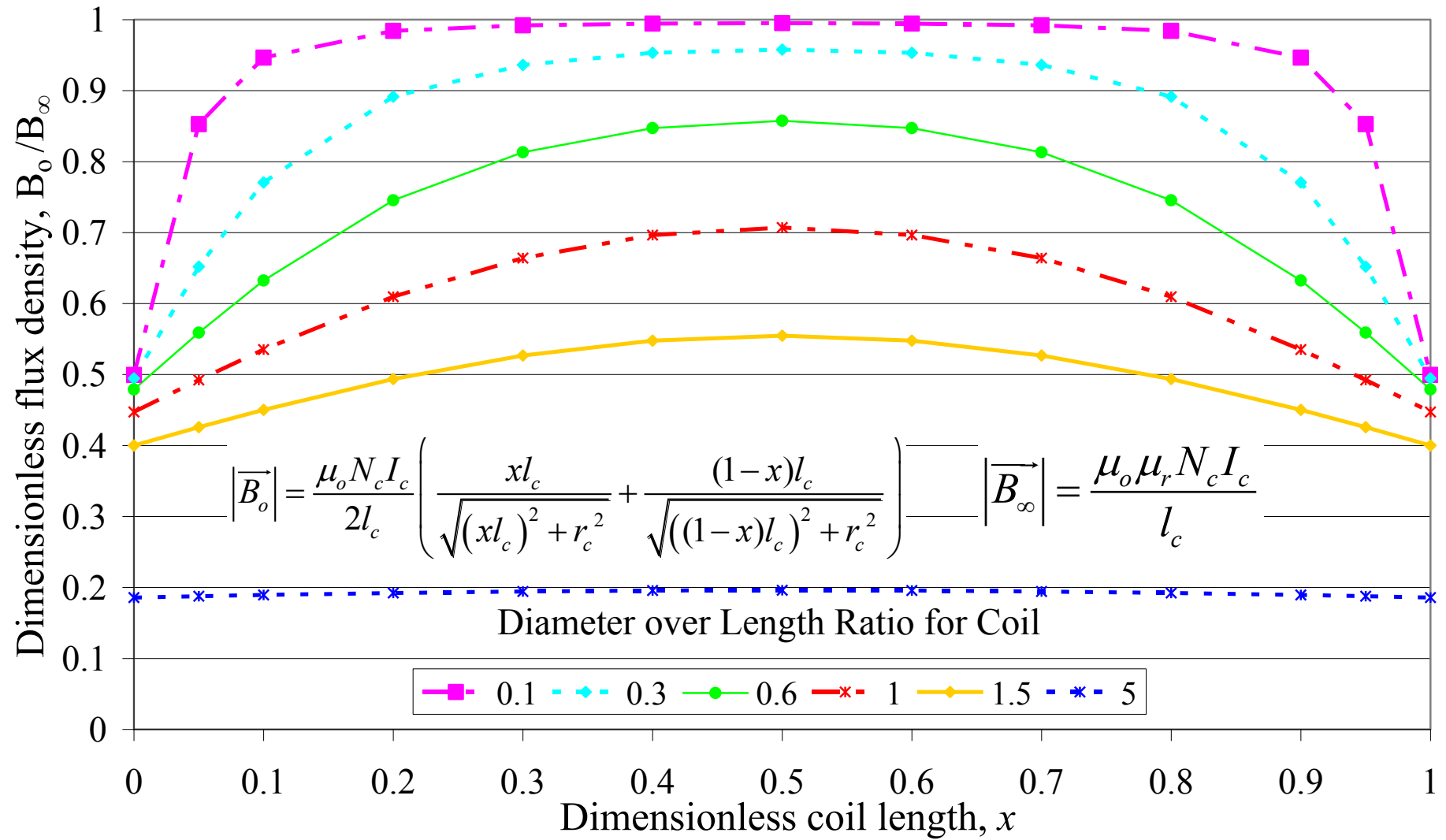
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Variation of magnetic penetration depth with frequency



- - A356 aluminium, 20 Deg. C

Centreline magnetic fields of short coils



Equations for 1D analytical and 2D axial symmetric model

$$P_w = k_N^{*2} \sqrt{2} \pi (I_c N_c)^2 \rho_w \xi_w \varphi(\xi_w) / l_c$$

$$k_N^* = k_N \left(1 - \left(\frac{D_w - \delta_w}{D_c + \delta_c} \right)^2 \right) + \left(\frac{D_w - \delta_w}{D_c + \delta_c} \right)^2$$

$$\varphi(\xi_w) = \frac{\sqrt{2} (\text{ber} \xi_w \text{ber}^{**} \xi_w + \text{bei} \xi_w \text{bei}^{**} \xi_w)}{\text{ber}^2(\xi_w) + \text{bei}^2(\xi_w)}$$

$$\xi_w = \frac{D_w}{\delta_w \sqrt{2}} \quad \delta_w = \left(\frac{\rho_w}{\pi \mu_r \mu_o f} \right)^{0.5}$$

$$\vec{B} = \nabla \times \vec{A}$$

$$\vec{B} = \mu_o \mu_r \vec{H}$$

$$\nabla \times \vec{H} = \vec{J}$$

$$\vec{J} = \sigma \vec{E} + \vec{J}^e$$

$$\vec{E} = \nabla V - \frac{\partial \vec{A}}{\partial t}$$

$$j\omega\sigma \vec{A}_\phi + \nabla \times \vec{H} = \vec{J}^e$$

Factors for analytical solutions

